

Algorithms for Statistical Signal Processing



John G. Proakis • Charles M. Rader
Fuyun Ling • Chrysostomos L. Nikias
Marc Moonen • Ian K. Proudler

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ALGORITHMS FOR STATISTICAL SIGNAL PROCESSING

John G. Proakis

Charles M. Rader

Fuyun Ling

Chrysostomos L. Nikias

Marc Moonen

Ian K. Proudler



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Preface

The field of digital signal processing (DSP) has expanded rapidly over the past three decades. During the late sixties and seventies, we witnessed the development of the basic theory for digital filter design and the development of computationally efficient algorithms for evaluating the Fourier transform, convolution, and correlation. During the past two decades, we experienced an explosion in DSP applications spurred by significant advances in digital computer technology and integrated-circuit fabrication. In this period, the basic DSP theory has expanded to include parametric signal modeling, with applications to power spectrum estimation and system modeling, adaptive signal processing algorithms, multirate and multidimensional signal processing, and higher-order statistical methods for signal processing.

With the expansion of basic DSP theory and the rapid growth in applications (spurred by the development of fast and inexpensive digital signal processors), there is a growing interest in advanced courses in DSP covering a variety of topics. This book was written with the goal of satisfying, in part, the resulting need for textbooks covering these advanced topics.

Most of the material contained in this book was first published in 1992 by the Macmillan Publishing Company, in a book entitled *Advanced Digital Signal Processing* (which went out of print in 1997). This new book differs from the earlier publication by the inclusion of a new chapter (Chapter 7) on QRD-based fast adaptive filter algorithms, and the deletion of a chapter on multirate signal processing. The other chapters have remained essentially the same.

The major focus of this book is on algorithms for statistical signal processing. Chapter 2 treats computationally efficient algorithms for convolution and for the computation of the discrete Fourier transform. Chapter 3 treats linear prediction and optimum

Wiener filters; included in this chapter is a description of the Levinson-Durbin and Schur algorithms. Chapter 4 considers the filter design problem based on the least-squares method and describes several methods for solving least squares problems, including the Givens transformation, the Householder transformation, and singular-value decomposition. Chapter 5 treats single-channel adaptive filters based on the LMS algorithm and on recursive least-squares algorithms. Chapter 6 describes computationally efficient recursive least-squares algorithms for multichannel signals. Chapter 7 is focused on the uses of signal flow graphs for deriving computationally efficient adaptive filter algorithms based on the QR decomposition. Chapter 8 deals with power spectrum estimation, including both parametric and nonparametric methods. Chapter 9 describes the use of higher-order statistical methods for signal modeling and system identification.

Although the material in this book was written by six different authors, we have tried very hard to maintain common notation throughout the book. We believe we have succeeded in developing a coherent treatment of the major topics outlined in the preceding overview. Chapter 1 provides an introduction to selected basic DSP material that is typically found in a first-level DSP text, and also serves to establish some of the notation used throughout the book.

In our treatment of the various topics covered herein, we generally assume that the reader has had a prior course on the fundamentals of digital signal processing. The fundamental topics assumed as background include the z -transform, the analysis and characterization of discrete-time systems, the Fourier transform, the discrete Fourier transform (DFT), and the design of FIR and IIR digital filters.

John G. Proakis
Charles M. Rader
Fuyun Ling
Chrysostomos L. Nikias
Marc Moonen
Ian K. Proudler

Contents

PREFACE	<i>xi</i>
1 INTRODUCTION	1
1.1 Characterization of Signals 2	
1.1.1 <i>Deterministic Signals, 2</i>	
1.1.2 <i>Random Signals, Correlation Functions, and Power Spectra, 5</i>	
1.2 Characterization of Linear Time-Invariant Systems 14	
1.2.1 <i>Time-Domain Characterization, 14</i>	
1.2.2 <i>Frequency-Domain Characterization, 17</i>	
1.2.3 <i>Causality and Stability, 19</i>	
1.2.4 <i>Bandpass Systems and Signals, 20</i>	
1.2.5 <i>Inverse Systems, Minimum-Phase Systems, and All-Pass Systems, 26</i>	
1.2.6 <i>Response of Linear Systems to Random Input Signals, 27</i>	
1.3 Sampling of Signals 30	
1.3.1 <i>Time-Domain Sampling of Analog Signals, 31</i>	
1.3.2 <i>Sampling the Spectrum of a Discrete-Time Signal, 38</i>	
1.3.3 <i>The Discrete Fourier Transform for Finite-Duration Sequences, 41</i>	
1.3.4 <i>The DFT and IDFT as Matrix Transformations, 43</i>	
1.4 Linear Filtering Methods Based on the DFT 46	
1.4.1 <i>Use of the DFT in Linear Filtering, 47</i>	
1.4.2 <i>Filtering of Long Data Sequences, 50</i>	

1.5	The Cepstrum	53
1.6	Summary and References	56
	Problems	56

2 ALGORITHMS FOR CONVOLUTION AND DFT

61

2.1	Modulo Polynomials	61
2.2	Circular Convolution as Polynomial Multiplication mod $u^N - 1$	63
2.3	A Continued Fraction of Polynomials	64
2.4	Chinese Remainder Theorem for Polynomials	66
2.5	Algorithms for Short Circular Convolutions	67
2.6	How We Count Multiplications	74
2.7	Cyclotomic Polynomials	76
2.8	Elementary Number Theory	77
	2.8.1 Greatest Common Divisors and Euler's Totient Function,	78
	2.8.2 The Equation $ax + by = 1$,	78
	2.8.3 Modulo Arithmetic,	81
	2.8.4 The Sino Representation of Integers Modulo M ,	83
	2.8.5 Exponentials Modulo M ,	85
2.9	Convolution Length and Dimension	88
2.10	The DFT as a Circular Convolution	92
2.11	Winograd's DFT Algorithm	95
2.12	Number-Theoretic Analogy of DFT	98
2.13	Number-Theoretic Transform	100
	2.13.1 Mersenne Number Transform,	104
	2.13.2 Fermat Number Transform,	106
	2.13.3 Considerations for Use of NTTs to Perform Circular Convolution,	107
	2.13.4 Use of Surrogate Fields for Complex Arithmetic,	108
2.14	Split-Radix FFT	110
2.15	Autogen Technique	116
2.16	Summary	122
	Problems	123

3 LINEAR PREDICTION AND OPTIMUM LINEAR FILTERS 125

- 3.1 Innovations Representation of a Stationary Random Process 125
 - 3.1.1 *Rational Power Spectra, 128*
 - 3.1.2 *Relationships between the Filter Parameters and the Autocorrelation Sequence, 129*
- 3.2 Forward and Backward Linear Prediction 131
 - 3.2.1 *Forward Linear Prediction, 131*
 - 3.2.2 *Backward Linear Prediction, 135*
 - 3.2.3 *Optimum Reflection Coefficients for the Lattice Forward and Backward Predictors, 139*
 - 3.2.4 *Relationship of an AR Process to Linear Prediction, 139*
- 3.3 Solution of the Normal Equations 140
 - 3.3.1 *Levinson-Durbin Algorithm, 140*
 - 3.3.2 *The Schur Algorithm, 144*
- 3.4 Properties of the Linear Prediction-Error Filters 148
- 3.5 AR Lattice and ARMA Lattice-Ladder Filters 152
 - 3.5.1 *AR Lattice Structure, 152*
 - 3.5.2 *ARMA Processes and Lattice-Ladder Filters, 154*
- 3.6 Wiener Filters for Filtering and Prediction 157
 - 3.6.1 *FIR Wiener Filter, 157*
 - 3.6.2 *Orthogonality Principle in Linear Mean-Square Estimation, 160*
 - 3.6.3 *IIR Wiener Filter, 161*
 - 3.6.4 *Noncausal Wiener Filter, 165*
- 3.7 Summary and References 167
- Problems 168

4 LEAST-SQUARES METHODS FOR SYSTEM MODELING AND FILTER DESIGN 177

- 4.1 System Modeling and Identification 178
 - 4.1.1 *System Identification Based on FIR (MA) System Model, 178*
 - 4.1.2 *System Identification Based on All-Pole (AR) System Model, 181*
 - 4.1.3 *System Identification Based on Pole-Zero (ARMA) System Model, 183*
- 4.2 Least-Squares Filter Design for Prediction and Deconvolution 189
 - 4.2.1 *Least-Squares Linear Prediction Filter, 190*
 - 4.2.2 *FIR Least-Squares Inverse Filters, 191*
 - 4.2.3 *Predictive Deconvolution, 195*
- 4.3 Solution of Least-Squares Estimation Problems 197
 - 4.3.1 *Definition and Basic Concepts, 198*
 - 4.3.2 *Matrix Formulation of Least-Squares Estimation, 199*

- 4.3.3 *Cholesky Decomposition*, 203
- 4.3.4 *LDU Decomposition*, 205
- 4.3.5 *QR Decomposition*, 207
- 4.3.6 *Gram-Schmidt Orthogonalization*, 209
- 4.3.7 *Givens Rotation*, 211
- 4.3.8 *Householder Reflection*, 214
- 4.3.9 *Singular-Value Decomposition*, 217
- 4.4 Summary and References 225
- Problems 226

5 ADAPTIVE FILTERS

231

- 5.1 Applications of Adaptive Filters 231
 - 5.1.1 *System Identification or System Modeling*, 233
 - 5.1.2 *Adaptive Channel Equalization*, 235
 - 5.1.3 *Echo Cancellation in Data Transmission over Telephone Channels*, 238
 - 5.1.4 *Suppression of Narrowband Interference in a Wideband Signal*, 242
 - 5.1.5 *Adaptive Line Enhancer*, 246
 - 5.1.6 *Adaptive Noise Cancelling*, 247
 - 5.1.7 *Linear Predictive Coding of Speech Signals*, 248
 - 5.1.8 *Adaptive Arrays*, 251
- 5.2 Adaptive Direct-Form FIR Filters 253
 - 5.2.1 *Minimum Mean-Square-Error Criterion*, 254
 - 5.2.2 *The LMS Algorithm*, 256
 - 5.2.3 *Properties of the LMS Algorithm*, 259
 - 5.2.4 *Recursive Least-Squares Algorithms for Direct-Form FIR Filters*, 265
 - 5.2.5 *Properties of the Direct-Form RLS Algorithms*, 273
- 5.3 Adaptive Lattice-Ladder Filters 276
 - 5.3.1 *Recursive Least-Squares Lattice-Ladder Algorithms*, 276
 - 5.3.2 *Gradient Lattice-Ladder Algorithm*, 300
 - 5.3.3 *Properties of Lattice-Ladder Algorithms*, 304
- 5.4 Summary and References 309
- Problems 309

6 RECURSIVE LEAST-SQUARES ALGORITHMS FOR ARRAY SIGNAL PROCESSING

314

- 6.1 QR Decomposition for Least-Squares Estimation 315
- 6.2 Gram-Schmidt Orthogonalization for Least-Squares Estimation 318
 - 6.2.1 *Least-Squares Estimation Using the MGS Algorithm*, 319
 - 6.2.2 *Physical Meaning of the Quantities in the MGS Algorithm*, 320
 - 6.2.3 *Time-Recursive Form of the Modified Gram-Schmidt Algorithm*, 321
 - 6.2.4 *Variations of the RMGS Algorithm*, 328

6.2.5	<i>Implementation of the RMGS Algorithm Using VLSI Arrays, and Its Relationship to the Least-Squares Lattice Algorithm, 332</i>	
6.3	Givens Algorithm for Time-Recursive Least-Squares Estimation 337	
6.3.1	<i>Time-Recursive Givens Algorithm, 337</i>	
6.3.2	<i>Givens Algorithm without Square Roots, 340</i>	
6.3.3	<i>The CORDIC Approach to Givens Transformations, 344</i>	
6.4	Recursive Least-Squares Estimation Based on the Householder Transformation 358	
6.4.1	<i>Block Time-Recursive Least-Squares Estimation Using the Householder Transformation, 358</i>	
6.5	Order-Recursive Least-Squares Estimation Algorithms 363	
6.5.1	<i>Fundamental Relations of ORLS Estimation, 364</i>	
6.5.2	<i>Canonical Structures for ORLS Estimation Algorithms, 370</i>	
6.5.3	<i>Variations in the Basic Processing Cells of ORLS Algorithms, 376</i>	
6.5.4	<i>Systematic Investigation and Derivation of ORLS Algorithms, 381</i>	
6.6	Summary and References 382	
	Problems 384	
7	QRD-BASED FAST ADAPTIVE FILTER ALGORITHMS	387
7.1	Background 388	
7.1.1	<i>Signal Flow Graphs, 388</i>	
7.1.2	<i>QRD-based RLS, Revisited, 390</i>	
7.1.3	<i>Residual Extraction, 392</i>	
7.2	QRD Lattice 394	
7.3	Multichannel Lattice 402	
7.4	Fast QR Algorithm 411	
7.5	Multichannel Fast QR Algorithm 416	
7.6	Summary and References 427	
	Problems 429	
8	POWER SPECTRUM ESTIMATION	432
8.1	Estimation of Spectra from Finite-Duration Observations of Signals 433	
8.1.1	<i>Computation of the Energy Density Spectrum, 433</i>	
8.1.2	<i>Estimation of the Autocorrelation and Power Spectrum of Random Signals: The Periodogram, 438</i>	
8.1.3	<i>Use of the DFT in Power Spectrum Estimation, 443</i>	

8.2	Nonparametric Methods for Power Spectrum Estimation	445
8.2.1	<i>Bartlett Method: Averaging Periodograms</i> ,	446
8.2.2	<i>Welch Method: Averaging Modified Periodograms</i> ,	447
8.2.3	<i>Blackman and Tukey Method: Smoothing the Periodogram</i> ,	449
8.2.4	<i>Performance Characteristics of Nonparametric Power Spectrum Estimators</i> ,	452
8.2.5	<i>Computational Requirements of Nonparametric Power Spectrum Estimates</i> ,	456
8.3	Parametric Methods for Power Spectrum Estimation	457
8.3.1	<i>Relationships Between the Autocorrelation and the Model Parameters</i> ,	459
8.3.2	<i>Yule-Walker Method for the AR Model Parameters</i> ,	461
8.3.3	<i>Burg Method for the AR Model Parameters</i> ,	462
8.3.4	<i>Unconstrained Least-Squares Method for the AR Model Parameters</i> ,	465
8.3.5	<i>Sequential Estimation Methods for the AR Model Parameters</i> ,	467
8.3.6	<i>Selection of AR Model Order</i> ,	468
8.3.7	<i>MA Model for Power Spectrum Estimation</i> ,	469
8.3.8	<i>ARMA Model for Power Spectrum Estimation</i> ,	470
8.3.9	<i>Experimental Results</i> ,	473
8.4	Minimum-Variance Spectral Estimation	481
8.5	Eigenanalysis Algorithms for Spectrum Estimation	483
8.5.1	<i>Pisarenko Harmonic Decomposition Method</i> ,	484
8.5.2	<i>Eigendecomposition of the Autocorrelation Matrix for Sinusoids in White Noise</i> ,	486
8.5.3	<i>MUSIC Algorithm</i> ,	488
8.5.4	<i>ESPRIT Algorithm</i> ,	489
8.5.5	<i>Order Selection Criteria</i> ,	492
8.5.6	<i>Experimental Results</i> ,	492
8.6	Summary and References	495
	Problems	496

9 SIGNAL ANALYSIS WITH HIGHER-ORDER SPECTRA

504

9.1	Use of Higher-Order Spectra in Signal Processing	504
9.2	Definition and Properties of Higher-Order Spectra	506
9.2.1	<i>Moments and Cumulants of Random Signals</i> ,	506
9.2.2	<i>Higher-Order Spectra (Cumulant Spectra)</i> ,	508
9.2.3	<i>Linear Non-Gaussian Processes</i> ,	510
9.2.4	<i>Nonlinear Processes</i> ,	512
9.3	Conventional Estimators for Higher-Order Spectra	514
9.3.1	<i>Indirect Method</i> ,	514
9.3.2	<i>Direct Method</i> ,	516

9.3.3	<i>Statistical Properties of Conventional Estimators</i> , 517	
9.3.4	<i>Test for Aliasing with the Bispectrum</i> , 518	
9.4	Parametric Methods for Higher-Order Spectrum Estimation	520
9.4.1	<i>MA Methods</i> , 522	
9.4.2	<i>Noncausal AR Methods</i> , 525	
9.4.3	<i>ARMA Methods</i> , 526	
9.4.4	<i>AR Methods for the Detection of Quadratic Phase Coupling</i> , 528	
9.5	Cepstra of Higher-Order Spectra	531
9.5.1	<i>Preliminaries</i> , 531	
9.5.2	<i>Complex and Differential Cepstra</i> , 532	
9.5.3	<i>Bicepstrum</i> , 533	
9.5.4	<i>Cepstrum of the Power Spectrum</i> , 535	
9.5.5	<i>Cepstrum of the Bicoherence</i> , 536	
9.5.6	<i>Summary of Cepstra and Key Observations</i> , 537	
9.6	Phase and Magnitude Retrieval from the Bispectrum	537
9.7	Summary and References	540
	Problems	541
	REFERENCES	542
	INDEX	559

Introduction

In this chapter, we review some basic topics in digital signal processing (DSP) and, in the process, establish notation that is used throughout the text. We begin with the characterization of deterministic and random signals in Section 1.1. In Section 1.2, we describe the characterization of linear time-invariant systems, in both the time and frequency domains. Included in this treatment are definitions of basic properties of systems, such as causality, stability, minimum phase, maximum phase, mixed phase, all-pass, and bandpass. The response of linear time-invariant systems to random input signals is also derived.

The third major topic of this chapter is concerned with the sampling of signals. Conditions are derived for alias-free sampling of continuous-time signals. Also treated in this section is the discrete Fourier transform (DFT) for finite duration sequences.

Linear filtering methods based on the use of the DFT make up the fourth major topic of this chapter. The final topic of the chapter is a description of the complex cepstrum of a signal. The use of the complex cepstrum in performing signal deconvolution is also treated briefly.

The foregoing topics are usually covered in a first course in digital signal processing. Consequently, our treatment is intended to serve as a brief review. Our choice of review topics was influenced by the advanced topics treated in this book; we should emphasize, however, that many other important topics have been omitted. We assume that the reader is familiar with z -transforms and Fourier transforms, and their use in the analysis of linear time-invariant systems. We also assume that the reader is familiar with filter design methods, and design tools and algorithms for both analog and digital filters. The introductory texts by Mitra (1998), Oppenheim and Schaffer (1989),

and Proakis and Manolakis (1996) provide the necessary background material for the topics treated in this book.

1.1 CHARACTERIZATION OF SIGNALS

A signal is defined as any physical quantity that varies with time, space, or any other independent variable or variables. Mathematically, we describe a signal as a function of one or more independent variables. If the signal is a function of a single independent variable, the signal is said to be *one-dimensional*. On the other hand, a signal is *M-dimensional* (multidimensional) if it is a function of M independent variables.

In some applications, signals are generated by multiple sources or multiple sensors. Such signals can be represented in vector form, where each element of the vector is a signal from a single source or a single sensor. The signal vector is called a *multichannel signal*.

In this book we deal mainly with one-dimensional, single-channel or multichannel signals for which the independent variable is time. When the independent variable is continuous, the signal is called a *continuous-time signal* or an analog signal. On the other hand, when the independent variable is discrete, the signal is called a *discrete-time signal*.

1.1.1 Deterministic Signals

Let us consider a deterministic continuous-time signal $x(t)$, which may be real- or complex-valued. We assume that the signal has finite energy, defined as

$$\mathcal{E} = \int_{-\infty}^{\infty} |x(t)|^2 dt \quad (1.1.1)$$

Such a signal is represented in the frequency domain by its Fourier transform

$$X(F) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi Ft} dt \quad (1.1.2)$$

where F is the frequency in cycles per second or hertz (Hz). From Parseval's theorem we have

$$\mathcal{E} = \int_{-\infty}^{\infty} |x(t)|^2 dt = \int_{-\infty}^{\infty} |X(F)|^2 dF \quad (1.1.3)$$

The quantity $|X(F)|^2$ represents the distribution of signal energy as a function of frequency and, hence, is called the *energy density spectrum*. It is denoted as

$$S_{xx}(F) = |X(F)|^2 \quad (1.1.4)$$

$S_{xx}(F)$ may also be viewed as the Fourier transform of another function, $r_{xx}(\tau)$, called the *autocorrelation function* of the finite energy signal $x(t)$, which is defined as

$$r_{xx}(\tau) = \int_{-\infty}^{\infty} x^*(t) x(t + \tau) dt \quad (1.1.5)$$

Indeed, it easily follows that

$$S_{xx}(F) = \int_{-\infty}^{\infty} r_{xx}(\tau) e^{-j2\pi F\tau} d\tau \quad (1.1.6)$$

so that $S_{xx}(F)$ and $r_{xx}(\tau)$ are a Fourier transform pair.

Similar relationships hold for discrete-time signals, which often are the result of uniformly sampling continuous-time signals. To be specific, suppose that $x(n)$ is a real or complex-valued sequence, where n takes integer values. If $x(n)$ is deterministic and has finite energy, that is,

$$\mathcal{E} = \sum_{n=-\infty}^{\infty} |x(n)|^2 \quad (1.1.7)$$

then $x(n)$ has the frequency domain representation

$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n} \quad (1.1.8)$$

or, equivalently,

$$X(f) = \sum_{n=-\infty}^{\infty} x(n) e^{-j2\pi f n} \quad (1.1.9)$$

where $\omega = 2\pi f$. The units for the frequency variables ω and f are radians and cycles, respectively. (Or, they are radians per sample interval and cycles per sample interval, if the sequence $x(n)$ is obtained by sampling a continuous-time signal $x(t)$ at a rate of $F_s = 1/T$ samples per second, where T is the sample interval. Then, $\omega = \Omega T = 2\pi FT$ and $f = FT$).

We note that $X(\omega)$ is periodic with period $\omega_p = 2\pi$ and $X(f)$ is periodic with period $f_p = 1$. In fact, the Fourier transform relationship in (1.1.9) may be interpreted as a Fourier series representation of the periodic function $X(f)$, where the sequence $\{x(n)\}$ constitutes the set of Fourier coefficients. Thus,

$$\begin{aligned} x(n) &= \int_{-1/2}^{1/2} X(f) e^{j2\pi f n} df \\ &= \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) e^{j\omega n} d\omega \end{aligned} \quad (1.1.10)$$

This relationship may also be viewed as the inverse Fourier transform that yields the sequence $x(n)$ from $X(f)$ or, equivalently, from $X(\omega)$.

By applying Parseval's theorem, the energy of the discrete-time sequence is also given as

$$\mathcal{E} = \int_{-1/2}^{1/2} |X(f)|^2 df \quad (1.1.11)$$

The quantity $|X(f)|^2$ represents the distribution of signal energy as a function of frequency and, hence, is called the *energy density spectrum* of the discrete-time signal.

It is denoted as

$$S_{xx}(f) = |X(f)|^2 \quad (1.1.12)$$

The energy density spectrum $S_{xx}(f)$ is related to the autocorrelation sequence

$$r_{xx}(m) = \sum_{n=-\infty}^{\infty} x^*(n)x(n+m) \quad (1.1.13)$$

via the Fourier transform. That is,

$$S_{xx}(f) = \sum_{m=-\infty}^{\infty} r_{xx}(m)e^{-j2\pi fm} \quad (1.1.14)$$

Two elementary deterministic signals that we will use frequently are the unit impulse and the unit step functions. In the continuous-time domain, the unit impulse may be defined by the property

$$\int_{-\infty}^{\infty} \delta(t)g(t) dt = g(0) \quad (1.1.15)$$

where $g(t)$ is an arbitrary function continuous at $t = 0$. Hence, its area is

$$\int_{-\infty}^{\infty} \delta(t) dt = 1 \quad (1.1.16)$$

The unit step function is defined as

$$u(t) = \begin{cases} 1, & t \geq 0 \\ 0, & t < 0 \end{cases} \quad (1.1.17)$$

In discrete-time, the unit sample, or unit impulse sequence, is defined as

$$\delta(n) = \begin{cases} 1, & n = 0 \\ 0, & n \neq 0 \end{cases} \quad (1.1.18)$$

The unit step sequence is denoted as $u(n)$ and defined as

$$u(n) = \begin{cases} 1, & n \geq 0 \\ 0, & n < 0 \end{cases} \quad (1.1.19)$$

A continuous-time signal $x(t)$ may be represented in general as the convolution of itself with a unit impulse,

$$x(t) = \int_{-\infty}^{\infty} x(\tau)\delta(t-\tau) d\tau \quad (1.1.20)$$

for all t . Similarly, a sequence $x(n)$ may be represented as

$$x(n) = \sum_{k=-\infty}^{\infty} x(k)\delta(n-k) \quad (1.1.21)$$

The expression in (1.1.21) is basically a convolution of the sequence $x(n)$ with the unit sample sequence $\delta(n)$. Equivalently, (1.1.21) may be viewed as the superposition (sum

over k) of unit sample sequences $\delta(n - k)$, scaled in amplitude by the corresponding values $x(k)$ of the sequence $x(n)$.

1.1.2 Random Signals, Correlation Functions, and Power Spectra

In this section we provide a brief review of the characterization of random signals in terms of statistical averages expressed in both the time and frequency domains. The reader is assumed to have a background in probability theory and random processes at least at the level given in the texts of Stark and Woods (1994), Leon-Garcia (1994), Helstrom (1991), Peebles (1987), Papoulis (1984), and Davenport (1970).

Random Processes. Many physical phenomena encountered in nature are best characterized in statistical terms. For example, meteorological phenomena such as air temperature and air pressure fluctuate randomly as a function of time. Thermal noise voltages generated in the resistors of an electronic device, such as a radio or television receiver, are also randomly fluctuating phenomena. These are just a few examples of random signals. Such signals are usually modeled as infinite-duration, infinite-energy signals.

Suppose that we take the set of waveforms corresponding to the air temperatures in different cities around the world. For each city there is a corresponding waveform that is a function of time, as illustrated in Fig. 1.1. The set of all possible waveforms is called an *ensemble* of time functions or, equivalently, a *random process*. The waveform for the temperature in any particular city is a *single realization* or a *sample function* of the random process. Similarly, the thermal noise voltage generated in a resistor is a single realization or a sample function of the random process that consists of all noise voltage waveforms generated by the set of all resistors.

The set (ensemble) of all possible waveforms of a random process is denoted as $X(t, S)$, where t represents the time index and S represents the set (sample space) of all possible sample functions. A single waveform in the ensemble is denoted by $x(t, s)$. Usually, we drop the variable s (or S) for notational convenience, so that the random process is denoted as $X(t)$ and a single realization is denoted as $x(t)$.

Having defined a random process $X(t)$ as an ensemble of sample functions, let us consider the values of the process for any set of time instants $t_1 > t_2 > t_3 > \cdots > t_n$, where n is any positive integer. In general, the samples $X_{t_i} \equiv X(t_i)$, $i = 1, 2, \dots, n$, are n random variables characterized statistically by their joint probability density function (pdf), denoted as $p(x_{t_1}, x_{t_2}, \dots, x_{t_n})$, and any n .

Stationary Random Process. Suppose that we have n samples of the random process $X(t)$ at $t = t_i$, $i = 1, 2, \dots, n$, and another set of n samples displaced in time from the first set by an amount τ . Thus the second set of samples are $X_{t_i+\tau} = X(t_i + \tau)$, $i = 1, 2, \dots, n$, as shown in Fig. 1.1. This second set of n random variables is characterized by the joint probability density function $p(x_{t_1+\tau}, \dots, x_{t_n+\tau})$. The joint pdf's of the two sets of random variables may or may not be identical. When they are