

**Analytical Techniques in the  
Theory of Guided Waves**

**R. Mittra and S. W. Lee**

# Analytical Techniques in the Theory of Guided Waves

R. Mittra and S. W. Lee

*University of Illinois*

THE MACMILLAN COMPANY · New York  
Collier-Macmillan Limited · London

COPYRIGHT © 1971, THE MACMILLAN COMPANY

PRINTED IN THE UNITED STATES OF AMERICA

All rights reserved. No part of this book may be reproduced or transmitted in any form or by any means, electronic or mechanical, including photocopying, recording, or any information storage and retrieval system, without permission in writing from the Publisher.

THE MACMILLAN COMPANY  
866 Third Avenue, New York, New York 10022

COLLIER-MACMILLAN CANADA, LTD., Toronto, Ontario

*Library of Congress catalog card number: 70-116784*

First Printing

# **Analytical Techniques in the Theory of Guided Waves**

**MACMILLAN SERIES IN ELECTRICAL SCIENCE**

**Roger F. Harrington, Editor**

# Preface

There has been considerable progress recently toward the development of analytical techniques for attacking boundary-value problems in guided-wave theory. This book is an attempt to present a unified account of a number of these techniques. It is intended to serve as a graduate-level text, and it is hoped that the book will also be found useful by researchers in the areas of electromagnetics and acoustics. The emphasis is on elaborating the principles of various mathematical techniques rather than on solving a large number of specific problems. Thus the same geometrical configuration is frequently chosen to serve as a typical example for the application of more than one analytical technique that may be employed to attack a problem. This permits convenient comparison of both the methods and the formats of solutions derived by these techniques. A large number of exercises have been included, accompanied by hints for solving them. These exercises deal with physical phenomena associated with the study of waveguide discontinuity, radiation and diffraction, and array problems and reference is made to the original papers that discuss these problems.

The book begins with the presentations of preliminary and background material in Chapter 1. Although some of this material is available in other texts, its inclusion is mainly for convenience of later reference. However, it is believed that Section 1-3, on the edge condition, and Section 1-4, on useful asymptotic formulas, contain materials that cannot be conveniently located in other sources.

Chapter 2 deals with the mode-matching technique, one of the most commonly used methods for formulating boundary-value problems in guided-wave theory. It illustrates the application of the direct-inversion and the residue-calculus methods for the exact solution of a class of problems involving waveguides and periodic structures. Chapters 2 and 3, both of which deal with classes of problems that possess exact solutions, serve as important background material for discussion of more advanced techniques in Chapters 4 and 5—techniques that are concerned with the derivation of semirigorous solutions to a much wider class of problems, which do not lend themselves to exact solution.

Chapter 3 presents a rather comprehensive discussion of the Wiener–Hopf technique based on the application of Fourier transforms and the theory of analytic continuation in complex variable theory. The Wiener–Hopf technique is typically discussed only very briefly, or in passing, in many texts on electromagnetic theory, although the technique provides a powerful tool for solving

a host of boundary-value problems that conform to a special type of geometry. The style of presentation of this material is somewhat different from that of Noble's text on Wiener-Hopf technique, although our Chapter 3 makes frequent use of the material appearing in that excellent book. Chapter 3 also includes some new forms of factorization formulas which enhance the usefulness of the Wiener-Hopf method. An additional feature is the discussion that establishes the connection between the Wiener-Hopf and the mode-matching techniques.

Chapters 4 and 5 are concerned with the generalization of the mode-matching and Wiener-Hopf techniques, with a view to significantly broadening their ranges of applications. The modified residue-calculus method discussed in Chapter 4 was developed only very recently but has found many applications to problems related to open and closed region waveguide discontinuity problems, phased arrays, and other periodic structures, particularly when applied in conjunction with the generalized scattering-matrix technique also discussed in this chapter. Another contribution of Chapter 4 is a description of the generalization of the mode-matching technique to open-region problems.

Finally, the Wiener-Hopf technique is generalized in Chapter 5 so as to be useful for a wider class of geometries. The generalization includes a case where the filling medium in the waveguide is inhomogeneous, a situation that has not been previously discussed elsewhere using the Wiener-Hopf technique.

Two other methods, the variational and quasi-static techniques, have been omitted from the list of topics because extensive discussions are readily available in a number of texts and reference books.

During the course of preparation of this book, the authors received encouragement and helpful criticism from many colleagues and friends at the University of Illinois, Professor G. A. Deschamps, Professor Y. T. Lo, and Mr. T. S. Li in particular. The book draws heavily upon the research publications and dissertations of former research students at the University of Illinois: Drs. J. R. Pace, D. S. Karjala, C. P. Bates, G. F. VanBlaricum, Jr., T. Itoh, and others. To them, the authors are deeply indebted. The expert secretarial help of Mrs. Lilian Beck, Mrs. Sharon Gocking, Mrs. Avi Oppenheim, and Mrs. Angie Johnson was much appreciated during the several stages of the preparation of the manuscript as the text evolved over a period of two years. Finally, much of the research work included in the text was sponsored by the Air Force Cambridge Research Laboratories under the monitorship of Mr. F. Zucker and Dr. R. A. Shore. The authors take this opportunity to express their thanks for the financial assistance received from the AFCRL through contract support.

*Urbana, Illinois*

R. M. and S. W. L.



# Basic Conventions and Notations

1. MKS units and  $e^{-i\omega t}$  time variation are used throughout (Section 1-2).
2.  $\psi(x, z)$  usually may be identified by  $H_y(x, z)$ , and  $\phi(x, z)$  by  $E_y(x, z)$  (Section 1-5).
3.  $\gamma = (\alpha^2 - k^2)^{1/2} = -i(k^2 - \alpha^2)^{1/2}$ , where  $\alpha = \sigma + i\tau$  and  $k = k_1 + ik_2$  (Section 1-6).
4. The Fourier transform pair is (Section 3-2)

$$\Psi(x, \alpha) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} \psi(x, z) e^{i\alpha z} dz$$
$$\psi(x, z) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} \Psi(x, \alpha) e^{-i\alpha z} d\alpha$$

5.  $C = \text{Euler's constant} = 0.57721 \dots$  (Section 1-4).



# **Analytical Techniques in the Theory of Guided Waves**

# Contents

Basic Conventions and Notations	ix
<b>Chapter 1. Preliminaries</b>	<b>1</b>
1-1 Introduction	1
1-2 Maxwell's Equations	1
1-3 Radiation Condition and Edge Condition	4
1-4 Some Formulas for Asymptotic Expansion	11
1-5 Cylindrical Waveguides and Periodic Structures	15
1-6 Modal Representation for Fields in an Open Region	20
1-7 Saddle-Point Method of Integration	23
<b>Chapter 2. Mode-Matching Techniques</b>	<b>30</b>
2-1 Introduction	30
2-2 Bifurcated Waveguide I: Formulation	30
2-3 Bifurcated Waveguide II: Solution by the Direct-Inversion Method	35
2-4 Bifurcated Waveguide III: Discussion of Solution	42
2-5 Bifurcated Waveguide IV: Solution by the Residue-Calculus Method	45
2-6 Radiation from an Infinite Array of Waveguides with Periodic Excitation	50
2-7 Radiation from an Infinite Array of Waveguides with Aperiodic Excitation	54
PROBLEMS	61
<b>Chapter 3. Wiener-Hopf Techniques</b>	<b>73</b>
3-1 Introduction	73
3-2 Fourier Transform	74
3-3 Properties of Fourier Transform in the Complex $\alpha$ -Plane	76
3-4 Integral-Equation Formulation	80
3-5 Solution of the Wiener-Hopf Equation	82
3-6 Bifurcated Waveguide	84
3-7 Jones's Method of Formulation	97

3-8	Radiation from an Infinite Array of Waveguides with Periodic Excitation	9
3-9	General Factorization and Decomposition Formulas	10
3-10	Alternative Factorization Formulas	11
3-11	Radiation from an Open-Ended Waveguide	12
3-12	Diffraction by an Open-Ended Waveguide	13
3-13	Mode-Matching and Wiener-Hopf Techniques	15
	PROBLEMS	15
<b>Chapter 4.</b>	<b>Generalization of the Mode-Matching Technique</b>	<b>16</b>
4-1	Introduction	16
4-2	Determinant Expansion Method	16
4-3	Modified Residue-Calculus Method	16
4-4	Mode Matching in the Fourier Transform Domain	18
4-5	Mode-Matching Technique for Open-Region Problems	19
4-6	Generalized Scattering-Matrix Technique	20
	PROBLEMS	21
<b>Chapter 5.</b>	<b>Approximate Methods Related to the Wiener-Hopf Technique</b>	<b>22</b>
5-1	Introduction	22
5-2	Semi-infinite Solid Cylindrical Antenna	22
5-3	Bifurcated Waveguide with Dielectric Loading	24
5-4	Slit-Coupled Parallel-Plate Waveguide	25
5-5	Diffraction by a Finite Waveguide	26
5-6	Radiation from Flanged Waveguide	27
	PROBLEMS	27
<b>Appendix</b>	<b>Proof of the Convergence of the Neumann Series Expansion</b>	<b>28</b>
	<b>Bibliography</b>	<b>29</b>
<b>Index</b>		<b>29</b>

# CHAPTER 1

## Preliminaries

### 1-1 Introduction

The purpose of this book is to give a unified account of a number of analytical methods that are found useful for solving a large class of open and closed waveguide problems. As a preparatory step we will present in this chapter some basic concepts and formulas to which subsequent discussions will frequently refer. The material included in Sections 1-2, 1-5, 1-6, and 1-7 can be found in several graduate-level textbooks and therefore serves only as a brief reminder. However, some of the discussion on edge condition, presented in a self-contained manner in Sections 1-3 and 1-4, may not be as readily accessible from other sources. The advanced reader may choose to bypass this chapter entirely and return to it when a particular reference is necessary.

### 1-2 Maxwell's Equations

The behavior of the electromagnetic fields in a continuous medium, isotropic or anisotropic, homogeneous or inhomogeneous, is governed by Maxwell's equations. In the MKS system of units, the differential forms of Maxwell's equations are

$$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t} \quad (2.1)$$

$$\nabla \times \mathbf{H} = \frac{\partial \mathbf{D}}{\partial t} + \mathbf{J} \quad (2.2)$$

$$\nabla \cdot \mathbf{B} = 0 \quad (2.3)$$

$$\nabla \cdot \mathbf{D} = \rho \quad (2.4)$$

$$\nabla \cdot \mathbf{J} = -\frac{\partial \rho}{\partial t} \quad (2.5)$$

with the various quantities defined as

$\mathbf{E}$ : *electric field* (in volts per meter)

$\mathbf{H}$ : *magnetic field* (in amperes per meter)

$\mathbf{D}$ : *electric flux density* (in coulombs per square meter)

$\mathbf{B}$ : *magnetic flux density* (in webers per square meter)

**J**: electric current density (in amperes per square meter)

$\bar{\rho}$ : electric charge density (in coulombs per cubic meter)

It is to be noted that these five equations, (2.1) through (2.5), are not independent. For example, with an appropriate initial condition we can obtain (2.3) by taking the divergence of (2.1). In a similar manner (2.4) may be derived from (2.2), in conjunction with (2.5) and appropriate initial conditions.

Throughout this book we will be concerned with *time-harmonic electromagnetic fields* only, and we will assume that all field quantities have a time variation given by  $\exp(-i\omega t)$ , where  $\omega$  is the angular frequency in radians.

Under this assumption the time derivatives in Maxwell's equations may be replaced by the factor  $-i\omega$ , and the common factor  $\exp(-i\omega t)$  may be dropped from these equations. Equations (2.1) through (2.5) then become

$$\nabla \times \mathbf{E} = i\omega \mathbf{B} \quad (2.6)$$

$$\nabla \times \mathbf{H} = -i\omega \mathbf{D} + \mathbf{J} \quad (2.7)$$

$$\nabla \cdot \mathbf{B} = 0 \quad (2.8)$$

$$\nabla \cdot \mathbf{D} = \rho \quad (2.9)$$

$$\nabla \cdot \mathbf{J} = i\omega \rho \quad (2.10)$$

Note that we have used boldface roman letters for the vectors that are complex functions of space coordinates only.

The set of Maxwell's equations given in (2.6) through (2.10) is not sufficient to determine the electromagnetic fields produced by given current and charge densities. The additional equations necessary for this purpose are supplied by relations between the fields (**E**, **H**), the flux densities (**D**, **B**), and the current density **J**, and are determined by the properties of the material medium involved.† These relations are generally known as the *constitutive relations* of the medium. The media are usually grouped into two categories: isotropic and anisotropic media.

*Isotropic media.* An isotropic medium is generally regarded as one in which the physical properties in the neighborhood of an interior point are the same in all directions. For most isotropic media, the constitutive relations are

$$\mathbf{D} = \epsilon \mathbf{E}, \quad \mathbf{B} = \mu \mathbf{H}, \quad \mathbf{J} = \sigma \mathbf{E} \quad (2.11)$$

If  $\epsilon$ ,  $\mu$ , and  $\sigma$  are not functions of position, the medium is said to be homogeneous; otherwise, it is called an inhomogeneous medium. The free space, or vacuum, is an isotropic medium in which

$$\epsilon = \epsilon_0 \approx \frac{1}{36\pi \times 10^9} \text{ farads/meter} \quad (2.12a)$$

---

† These relations are also dependent on the frames of reference of the medium and the observer if there is relative motion between these two frames.

$$\mu = \mu_0 = 4\pi \times 10^{-7} \text{ henrys/meter} \quad (2.12b)$$

$$\sigma = 0 \quad (2.12c)$$

For other isotropic media, it is convenient to introduce the dimensionless ratios

$$\epsilon_r = \frac{\epsilon}{\epsilon_0}, \quad \mu_r = \frac{\mu}{\mu_0} \quad (2.13)$$

which are generally labeled the relative dielectric constant and relative permeability, respectively.

*Anisotropic media.* In an anisotropic medium, the physical properties in the neighborhood of a point may be different for different directions. Typical examples are crystals, magnetized ferrites, and ionized media with externally applied static magnetic fields. Their constitutive relations can be generally represented by

$$\mathbf{D} = \boldsymbol{\epsilon} \cdot \mathbf{E}, \quad \mathbf{B} = \boldsymbol{\mu} \cdot \mathbf{H}, \quad \mathbf{J} = \boldsymbol{\sigma} \cdot \mathbf{E} \quad (2.14)$$

where  $\boldsymbol{\epsilon}$ ,  $\boldsymbol{\mu}$ , and  $\boldsymbol{\sigma}$  are tensors of rank two, alternatively termed dyadics. We will not deal with anisotropic media except in certain exercises, at which time we will present the explicit forms of  $\boldsymbol{\epsilon}$ ,  $\boldsymbol{\mu}$ , and  $\boldsymbol{\sigma}$  for some special anisotropic media.

In dealing with guided-wave problems, we often face a situation in which the physical properties of the medium change abruptly across one or several surfaces. The behavior of the fields in the presence of such discontinuities is governed by certain boundary conditions to be satisfied at the surfaces of the discontinuities. These conditions may be derived by an application of Maxwell's equations to infinitesimally small regions containing these surfaces. Some explicit forms of boundary conditions are as follows:

1. *At a material boundary* (discontinuous  $\epsilon$  and  $\mu$ ; refer to Figure 1-1). If the media in regions 1 and 2 have finite conductivities, the tangential electric and magnetic components are continuous across the boundary. That is,

$$\mathbf{n} \times (\mathbf{E}^{(2)} - \mathbf{E}^{(1)}) = 0 \quad (2.15a)$$

$$\mathbf{n} \times (\mathbf{H}^{(2)} - \mathbf{H}^{(1)}) = 0 \quad (2.15b)$$

where  $\mathbf{n}$  is the unit outward normal viewed from region 1.

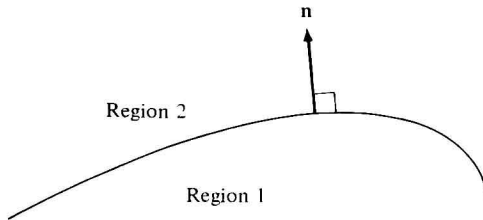


FIGURE 1-1 Boundary surface.

2. *At a perfectly conducting surface.* Let the medium in region 1 in Figure 1-1 be a perfect conductor (with an infinitely large conductivity  $\sigma$ ). Then the boundary conditions are

$$\mathbf{n} \times \mathbf{E}^{(2)} = 0 \quad (2.16)$$

$$\mathbf{n} \times (\mathbf{H}^{(2)} - \mathbf{H}^{(1)}) = \mathbf{J}_s \quad (2.16)$$

where  $\mathbf{J}_s$  is the surface current density.

### 1-3 Radiation Condition and Edge Condition

In certain situations in which the region of interest either involves boundaries at infinity or contains geometrical singularities, it is possible to derive several mathematically acceptable solutions of Maxwell's equations, only one of which is consistent with anticipated physical phenomenon. Therefore, in these situations it becomes necessary to apply certain additional physical constraints to ensure the uniqueness of the solutions.

In an unbounded space with all sources contained in a finite region, the additional constraint that governs the behavior of the fields at infinity is stated in terms of the *radiation condition*, which may be applied in one of two ways. If the medium in the space is lossy, we require that *the fields vanish at infinity*. If the medium is lossless and isotropic, the behavior of the fields at infinity is governed by the *Sommerfeld radiation condition*, which may be stated as follows. The field at a large distance  $r$  from the source has a phase progression outward and has an amplitude that decreases at least as rapidly as  $r^{-1}$ . More precisely, any transverse components  $\psi$  of the field (with respect to the direction) must satisfy the condition

$$\lim_{r \rightarrow \infty} r \left( \frac{\partial \psi}{\partial r} - ik\psi \right) = 0 \quad (3)$$

where  $k = \omega \sqrt{\mu\epsilon}$  is the propagation constant of the medium.

Yet another situation, where the solution of Maxwell's equations may not be unique, arises when the configuration of the problem contains geometrical singularities, such as sharp edges. The additional physical condition needed here, known as the *edge condition*, is supplied by the requirement that the electrical and magnetic energy stored in any finite neighborhood of the edge must be finite; that is,

$$\int_V (\epsilon |\mathbf{E}|^2 + \mu |\mathbf{H}|^2) dv \rightarrow 0 \quad (3)$$

as the volume  $V$  contracts to the neighborhood of the edge. For a smooth edge, which may be regarded as locally straight, the differential volume



(3.2) is  $dv = \rho \, d\rho \, d\phi \, dz$ , where  $(\rho, \phi, z)$  is the locally cylindrical coordinate of the edge. Then from (3.2) one may deduce that in the neighborhood of the edge, *none of the field components of  $(\mathbf{E}, \mathbf{H})$  should grow more rapidly than  $\rho^{-1+\tau}$  with  $\tau > 0$  as  $\rho \rightarrow 0$* . Strictly speaking, it is not necessary to know a priori the exact value of  $\tau$  but only its lower bound, which is greater than zero, in order to derive a unique solution to Maxwell's equations. In many instances, however, it is convenient to have a prior knowledge of  $\tau$ . We will now illustrate how the characteristic value  $\tau$  can be calculated from Maxwell's equations and the knowledge of the edge configuration. The method we follow is based on a study by Meixner.<sup>†</sup> For problems encountered in this book, it will be sufficient to consider, for the purpose of determining  $\tau$ , a two-dimensional perfectly conducting wedge as shown in Figure 1-2. The three media surrounding the edge in regions 1, 2, and 3 are characterized by  $(\mu_1, \epsilon_1)$ ,  $(\mu_2, \epsilon_2)$ , and

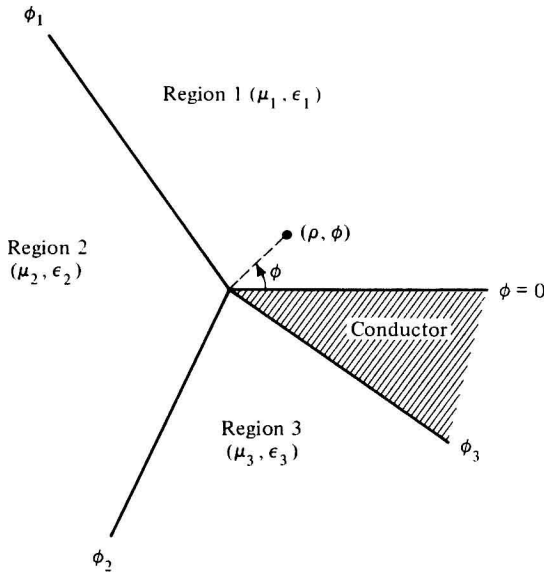


FIGURE 1-2 Perfectly conducting wedge surrounded by three different isotropic media.

$(\mu_3, \epsilon_3)$ , respectively. The angles,  $\phi_1$ ,  $\phi_2$ , and  $\phi_3$ , are all defined between 0 and  $2\pi$ . Maxwell's equations (2.6) and (2.7), together with the constitutive relations (2.11), may be written in the cylindrical coordinate system  $(\rho, \phi, z)$  as follows:

<sup>†</sup> J. Meixner, "The Behavior of Electromagnetic Fields at Edges," Inst. Math. Sci. Res. Rept. EM-72, New York University, New York, N.Y., Dec. 1954.

$$\begin{aligned}\frac{1}{\rho} \frac{\partial E_z}{\partial \phi} - \frac{\partial E_\phi}{\partial z} &= i\omega\mu H_\rho \\ \frac{\partial E_\rho}{\partial z} - \frac{\partial E_z}{\partial \rho} &= i\omega\mu H_\phi\end{aligned}\quad (3)$$

$$\begin{aligned}\frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho E_\phi) - \frac{1}{\rho} \frac{\partial E_\rho}{\partial \phi} &= i\omega\mu H_z \\ \frac{1}{\rho} \frac{\partial H_z}{\partial \phi} - \frac{\partial H_\phi}{\partial z} &= -i\omega\epsilon E_\rho \\ \frac{\partial H_\rho}{\partial z} - \frac{\partial H_z}{\partial \rho} &= -i\omega\epsilon E_\phi\end{aligned}\quad (3)$$

$$\frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho H_\phi) - \frac{1}{\rho} \frac{\partial H_\rho}{\partial \phi} = -i\omega\epsilon E_z$$

Since the field behavior near the edge  $\rho = 0$  is of interest, we may expand each of the field components in different angular regions as a power series in  $\rho$ . Recall from the edge condition that in the neighborhood of the edge components of the field can grow more rapidly than  $\rho^{-1+\tau}$ , with  $\tau > 0$ . With this in mind we may write

$$\begin{aligned}E_\rho &= \rho^{-1+\tau}[a_0^{(j)} + a_1^{(j)}\rho + a_2^{(j)}\rho^2 + \cdots] \\ E_\phi &= \rho^{-1+\tau}[b_0^{(j)} + b_1^{(j)}\rho + b_2^{(j)}\rho^2 + \cdots] \\ E_z &= \rho^{-1+\tau}[c_0^{(j)} + c_1^{(j)}\rho + c_2^{(j)}\rho^2 + \cdots] \\ H_\rho &= \rho^{-1+\tau}[A_0^{(j)} + A_1^{(j)}\rho + A_2^{(j)}\rho^2 + \cdots] \\ H_\phi &= \rho^{-1+\tau}[B_0^{(j)} + B_1^{(j)}\rho + B_2^{(j)}\rho^2 + \cdots] \\ H_z &= \rho^{-1+\tau}[C_0^{(j)} + C_1^{(j)}\rho + C_2^{(j)}\rho^2 + \cdots]\end{aligned}\quad (3)$$

where  $j = 1, 2$ , and  $3$ , corresponding to fields in the three regions of Fig. 1-2. The coefficients in (3.5) and (3.6) are functions of  $\phi$  and  $z$  only. Some of the relations between these coefficients can be determined by inserting (3.5) and (3.6) into Maxwell's equations in (3.3) and (3.4) and comparing the coefficients of equal powers of  $\rho$ . When this is done, the following relations are obtained:

$$c_0^{(j)}(\tau - 1) = 0 \quad (3.7)$$

$$-i\omega\epsilon_j a_0^{(j)} = \frac{\partial C_1^{(j)}}{\partial \phi} - \frac{\partial B_0^{(j)}}{\partial z} \quad (3.8)$$

$$-i\omega\epsilon_j b_0^{(j)} = \frac{\partial A_0^{(j)}}{\partial z} - \tau C_1^{(j)} \quad (3.9)$$

$$0 = \tau B_0^{(j)} - \frac{\partial A_0^{(j)}}{\partial \phi} \quad (3.10)$$