## low-noise electronic design

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AND F.C. FITCHEN

# Low-Noise Electronic Design

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### **PREFACE**

This book is intended for use by practicing engineers and by students of electronics. It can be used for self-study or in an organized classroom situation as a quarter or semester course or short course. A knowledge of electric circuit analysis, the principles of electronic circuits, and mathematics through an introduction to the calculus is assumed.

The approach used in this text is practice or design oriented. The material is not a study of noise, but rather of methods to deal with the everpresent noise in electronic systems.

Electrical noise is a problem in industrial, military, and consumer equipment design. The designer must be cognizant not only of the sources of noise, but also of the methods of noise reduction that are available to him. He will strive toward an optimum design. In this quest he must use all available and applicable tools. One such tool is the digital computer. In Chapter 8, computer-aided design of low-noise systems is considered. Another powerful tool is the laboratory. Chapter 14 discusses noise measurement methods and techniques.

The book is divided into two parts. Part I, Chapters I through 6, is primarily concerned with noise mechanisms, noise models, and the analysis of noisy circuits and systems. Part II, starting with Chapter 7, focuses on the design of low-noise circuits and systems. Many design examples are given.

The treatment may be considered to be oriented toward low- and midfrequency applications. The examples given are applicable to the instrumentation, audio, and control fields. Whereas there are few direct references to high-frequency communications, much of the material presented can be applied to that area of activity.

A summary of the most important points and topics is included at the end of each chapter. It is hoped that these summaries will be useful to the reader who does not have the time to start from the beginning. For the student who

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is encountering this material for the first time, the summaries serve to accentuate the more important aspects of the presentation. A controlled amount of repetition is included in the text, and many cross-references are made to other sections of the book for prerequisite or additional material. Problems are included at the end of each chapter and answers to the problems are given following Appendix IV.

This book is an outgrowth of our research, development, design, and teaching experiences. Special recognition is given to Honeywell, Incorporated, for cooperation and support.

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Hopkins, Minnesota Bridgeport, Connecticut July 1972

### **SYMBOLS**

Symbol	Definition	First used or Defined in Equation
$A_i$	Current gain	101
$A_v$	Voltage gain	1–5
$A_{vo}$	Midband voltage gain	1-5
BW	Bandwidth	2-27
C	Capacitance	1-16
C	Correlation coefficient	1-8
C	Coulombs, electric charge	1–12
$C_{b'c}, C_{b'e}$	Hybrid-π transistor model capacitances	4_4
$C_p$	Shunt capacitance	3-10
$\overline{CMV}$	Common-mode voltage	Ch. 11
$C_{\pi}$	Base-emitter capacitance $(C_{b'e})$	12-66
$E, E_1, E_2$ etc.	rms noise voltages	1–7
$E_A, E_B, \ldots, E_F$	rms noise voltage of resistors $R_A$ , $R_B$ ,, $R_F$	12-3
$E_b$	Thermal noise of $r_{bb}$	4-10
$E_f$	rms noise voltage with 1/f power spectrum	1-10
$E_n$	rms equivalent noise voltage generator	2-7
$E_{nb}$	rms narrowband noise voltage	2-24
$E_{ni}$	Equivalent input noise	2-1
$E_{no}$	rms noise at output of network	1-16
$E_{ns}$	rms noise of source resistance	3–3
$E_{nT}$	Total equivalent input noise voltage	12-73
$E_{n1}, E_{n2}, E_{n3}$	Noise voltage generators of stages 1, 2, 3	12-11
$E_{n1}, E_{n2}, E_{n3}$ $E_{s}$	rms noise of source resistance $R_s$	4–10
$E_{sh}$	Shot noise voltage of a diode	14–19

Symbol	Definition	First used or Defined in Equation
$E_t$	rms thermal noise voltage of a resistance or	
	conductance	1-3a
$E_{w^{\gamma}}$	rms wide band noise voltage	2-24
F	Noise factor	2-9
f	Frequency	1-10
l f	(read "one-over-f") noise source whose power level is inversely proportional to frequency	1-9
$f_{ m hfe}$	Beta-cutoff frequency = $f_B$	4–5
$f_{\hat{c}}$	High frequency 3 dB point	1-9
$f_t$	Lower frequency 3 dB point	1-9
$f_L^{t}$	Lower noise-corner frequency Hybrid- $\pi$	1-7
,	transistor model	4-8
$F_{\text{opt}}$	Optimum noise factor	2-14
$f_T$	Gain-bandwidth product	4–5
$G_a$	Available power gain	2–17
$g_m$	Transconductance	4-2
$G_a$	Peak power gain	1-4
$I_{\scriptscriptstyle B}$	de base current	4-14
$I_{h}$	Shot noise of base current	4-10
$I_{bb}$	Burst noise current generator	<del>7</del> -10 5-1
$I_{C}$	de collector current	4–14
$I_c$	Shot noise of collector current	4–10
$oldsymbol{I_{CBO}}$	Collector-base leakage current with emitter	4-10
*CBO	open	8-3
$I_{DC}$	Direct current	1–12
$I_{\scriptscriptstyle E}$	dc emitter current	1–13
$I_{ex}$	Excess 1/f noise current	3–11
$I_{f}$	I/f noise current in Hybrid-π model	4–8
$I_n$	rms equivalent noise current generator	2-7
$I_{no}$	rms noise current out of a network	410
$I_{nT}$	Total equivalent input noise current	12-74
$I_{n1}, I_{n2}, I_{n3}$	Noise current generators of stages 1, 2, 3	12-11
$I_0$	Saturated value of reverse diode current	1-13
$I_{sh}$	rms shot noise current	1-12
$I_t$	rms thermal noise current of a resistance	1–7
k	Boltzmann's constant $1.38 \times 10^{-23} \text{ W-sec/}^{\circ}\text{K}$	. 1–1
$K$ , $K_1$ , $K_2$ etc.	Arbitrary constants	1–9
$K_t$	System transfer voltage gain	2–3
$K_{tc}$	Transfer voltage gain of cascaded stages	12-16

		First used or Defined
Symbol	Definition	in Equation
$N_f$	Noise power of signal source with 1/f frequency	
	distribution	1-9
$N_i$	Input noise power	2-10
$N_o$	Output noise power	2–10
$N_o$	Available noise power	2–17
$N_t$	Available noise power	1-1
NEP	Noise equivalent power	3–14
NF	Noise figure	2-11
NI	Noise index in $\mu V/V/Decade$	9–2
Q	Quality factor of a tuned circuit	8–9
q	Electronic charge $6.02 \times 10^{-19}$ coulomb	1–12
R	Resistance, real part of Z	1–2
$r_b$	Base resistance for 1/f noise	4–9
$r_b$	Base resistance in T-equivalent transistor model	4–30
$r_{bb'}, r_{b'e},$ $r_{ce}, r_{b'c}$	Hybrid-π transistor model resistances	4–3
$R_C$	Collector resistor	10-4
$R_E$	Emitter resistor	10-10
$R_e$	Absolute value of emitter impedance	10-3
$r_e$	Schockley emitter resistance, emitter resistance	1–15
	in common-base T model	
$R_i$	Input resistance	10–6 10–1
$R_L$	Load resistance	
$R_n$	Equivalent noise resistance	2–15
$R_o$	Optimum source resistance	2-13
$R_p$	Parallel resistor	3–5
$R_{\rm s}$	Resistance of signal source	2-2
$r_x$	Base resistance $(r_{bb'})$	12-1
$r_0$	Output resistance ( $\simeq r_{ce}$ )	12-1
$r_{\pi}$	Base-emitter resistance $(r_{b'e})$	12-1
S/N	Signal-to-noise ratio	3–3
$S_i$	Input signal power	2–10
S(f)	Noise spectral density	1-6
$S_o$	Output signal power	2–10
T	Temperature in degrees Kelvin (°K)	1-1
T	Transformer turns ratio	7-3
$T_s$	Equivalent noise temperature	2–16
$V_{\scriptscriptstyle BE}$	Potential applied between base and emitter	1-13
$V_{cc}$	dc collector supply voltage	10-10

### xviii Symbols

Symbol	Definition	First used or Defined in Equation
$V_m$	Peak value of ac signal	11-1
$V_o$	Signal voltage at the output of a network	2-3
$V_{\rm s}$	Signal voltage at the source	2-3
$V_1$ , $V_2$ , etc.	Signal voltages	1-7
$X_C$	Reactance of a capacitance	1-16
$Z_i$	Input impedance	2–2
$Z_s$	Source impedance	7–5
α	Transistor current gain, T-equivalent circuit	4–32
$\beta_o$	Short-circuit current gain	4–1
$\Delta f$	Noise bandwidth	1-4
γ	Exponent in 1/f noise model	48
$\Lambda$	q/kT	4-2
w	$2\pi f$	1-16

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### **INTRODUCTION**

Sensors, detectors, and transducers are basic to the instrumentation and control fields. They are the "fingers" and "eyes" that reach out and measure. They must translate the characteristics of the physical world into electrical signals. We process and measure these signals and interpret them to be the reactions taking place in a chemical plant, the environment of an orbiting satellite, or the odor of an onion. An engineering problem often associated with sensing systems is the level of electrical noise generated in the sensor and in the electronic system.

In recent years, new high-resolution sensors and high-performance systems have been developed. All sensors have a basic or limiting noise level. The system designer must interface the sensor with electronic circuitry that contributes a minimum of additional noise. To raise the signal level an amplifier must be designed to complement the sensor. Achieving optimum system performance is a primary consideration in this book.

The following chapters answer several important questions. When we are given a sensor with specific impedance, signal and noise characteristics, how can we design or select an amplifier for minimum noise contribution, and concurrently maximize the signal-to-noise ratio of the system? If we have a sensor-amplifier system complete with known signal and noise, we must determine the major noise source; is it sensor, amplifier, or pickup noise? Are we maximizing the signal? Is improvement possible?

This book is a study of low-noise electronic design, not a treatise on noise as a physical phenomenon. It is divided into two parts. The first six chapters are concerned with *Noise Mechanisms and Models*; Part II, Chapters 7 through 13, deals with *Design Techniques and Examples*. The final chapter is a treatment of noise measurement methods.

In Part I we are concerned with the sources of noise; these sources include sensors and other devices, amplifiers, and associated circuitry. Noise models are developed for circuit components and for subsystems, and the relations between noise sources, biasing circuits, and operating point selection are examined. Analysis uses the noise models of system components to predict the expected minimum noise performance.

To realize the noise performance predicted from a system analysis many design decisions are required. Part II examines all the areas of possible noise

### 2 Introduction

problems, including component noise, noise in power supplies, shielding, biasing circuit noise, feedback, multistage configurations, as well as the associated gain, stability, and frequency response behavior. Design examples are given of noiseless biasing, low-noise power supplies, amplifier pairs, and complete low-noise systems.

#### **SYMBOLS**

To designate electric circuit quantities the system of symbols used in this book conforms to standard practice in the semiconductor field.\*

- 1. Dc values of quantities are indicated by capital letters with capital subscripts  $(I_B, V_{CE})$ . Direct supply voltages have repeated subscripts  $(V_{BB}, V_{CC})$ .
- 2. Rms values of quantities are indicated by capital letters with lowercase subscripts ( $I_c$ ,  $V_{ce}$ ).
- 3. Maximum or peak values are designated as rms values but bear an additional final subscript m ( $V_{\rm cm},\,V_{\rm bem}$ ).
- 4. The time-varying components of voltages and currents are designated by lowercase letters with lowercase subscripts  $(i_e, v_{ce})$ .
- 5. Instantaneous total values are represented by lowercase letters with capital subscripts  $(i_C, v_{CE})$ .

An example may be helpful. A time-varying waveform.

$$i_b = I_{bm} \sin \omega t$$

when added to a dc wave of average value  $I_B$ , is represented by its total value

$$i_B = I_B + i_b$$

So a complete expression is

$$i_B = I_B + I_{bm} \sin \omega t$$

The rms value of the time-varying portion of this composite wave is  $I_b$ .

The symbols for rms, peak, and instantaneous quantities (2, 3, and 4) apply only to the time-varying portion of a waveform. No symbols are proposed for the rms and peak values of a wave containing both time-varying and do components.

#### **PREFIXES**

The following prefixes are used to indicate decimal multiples or submultiples of units:

\* "IEEE Standard Letter Symbols for Semiconductor Devices," IEEE Trans. Electron Devices, ED-11, No. 8, August 1964.